AN1568/D

Interfacing Between LVDS and ECL

Prepared by: Paul Lee Logic Applications Engineer ON Semiconductor



http://onsemi.com

APPLICATION NOTE

Introduction

Recent growth in high-speed data transmission between high-speed ICs demand more bandwidth than ever before while still maintaining high performance, low power consumption and good noise immunity. Emitter Coupled Logic (ECL) recognized the challenge and provided high performance and good noise immune devices. ECL migrated toward low voltages to reduce the power consumption and to keep up with current technology trends by offering 3.3 V and 2.5 V Low Voltage ECL (LVECL) devices. LVECL devices working off the positive 3.3 V or 2.5 V supply voltages are called the Low Voltage Positive ECL (LVPECL). LVPECL maintains 750 mV output swing with a 0.9 V offset from V_{CC}, which makes them ideal as peripheral components.

LVDS (Low Voltage Differential Signaling) technology also addresses the needs of current high performance applications. LVDS as specified in ANSI/TIA/EIA-644 by Data Transmission Interface committee TR30.2 and IEEE 1596.3 SCI-LVDS by IEEE Scalable Coherent Interface standard (SCI) is a high speed, low power interface that is a solution in many application areas. LVDS provides an output swing of 250 mV to 400 mV with a DC offset of 1.2 V. External resistor components are required for board-to-board data transfer or clock distribution.

LVPECL and LVDS are both differential low voltage signals, but with different amplitude and different offset. The purpose of this documentation is to show the interfacing between ECL and LVDS. In addition, it gives interface suggestions to and from 5.0 V supplied PECL devices and negative supplied ECL or NECL

ECL levels

Today's applications typically use ECL devices in the PECL mode. PECL (Positive ECL) is nothing more than supplying any ECL device with a positive power supply (V_{CC} = +5.0 V, V_{EE} = 0 V). In addition, ECL uses differential data transmission technology, which results in better noise immunity. Since the common mode noise is coupled onto the differential interconnect, it will be seen as a common mode modulation and will be rejected.

With the trend towards low voltage systems, a new generation of ECL circuitry has been developed. The Low Voltage NECL (LVPECL) devices work using negative -3.3 V or -2.5 V power supply, or more popular positive power supplies, $V_{CC} = +3.3 \text{ V}$ or +2.5 V and $V_{EE} = \text{GND}$ as LVPECL.

The temperature compensated (100EL, 100LVEL, 100EP, 100LVEP) output DC levels for the different supply levels are shown in Table 1. ECL outputs are designed as an open emitter, requiring a DC path to a more negative supply than V_{OL} . A pull down resistor termination can be used to terminate the transmission line (see AND8020 for ECL Termination information).

ECL standard DC input levels are also relative to $V_{\rm CC}$. Many devices are available with Voltage Input HIGH Common Mode Range ($V_{\rm IHCMR}$). These differential inputs allow processing signals with small swings (down to 200 mV, 150 mV or even 50 mV signal levels) within an appropriate offset range. The $V_{\rm IHCMR}$ ranges of ECL devices are listed in each respective data sheets.

Table 1. MC100EXXX/MC100ELXXX/LVELXXX/EPXXX/LVEPXXX ($T_A = 0$ °C to +85°C)

Symbol	Parameter	2.5 V LVPECL (Note 1)	3.3 V LVPECL (Note 1)	5.0 V PECL (Note 1)	NECL	Unit
V _{CC}		+2.5	+3.3	+5	GND	V
V _{EE}		GND	GND	GND	-5.2, -4.5, -3.3 or -2.5	V
V _{OH}	Maximum Output HIGH Level	1.680	2.480	4.180	-0.820	V
V _{OH}	Typical Output HIGH Level	1.555	2.355	4.055	-0.945	V
V _{OH}	Minimum Output HIGH Level	1.430	2.230	3.930	-1.070	V
V _{OL}	Maximum Output LOW Level	0.880	1.680	3.380	-1.620	V
V _{OL}	Typical Output LOW Level	0.755	1.555	3.255	-1.745	V
V _{OL}	Minimum Output LOW Level	0.630	1.430	3.130	-1.870	V

^{1.} All levels vary 1:1 with V_{CC}

LVDS Levels

As the name indicates, the LVDS main attribute is the low voltage amplitude levels compared to other data transmission standards, as shown in Figure 1. The LVDS specification states 250 mV to 400 mV output swing for driver/transmitter (VPP). The low voltage swing levels result in low power consumption while maintaining high performance levels required by most users. In addition, LVDS uses differential data transmission technology equivalent to ECL. Furthermore, LVDS technology is not dependent on specific power supply levels like ECL technology. This signifies an easy migration path to lower supply voltages such as 3.3 V, 2.5 V, or lower voltages while still maintaining the same signaling levels and high performance. ON Semiconductor currently provides a 2.5 V 1:5 dual differential LVDS Clock Driver/Receiver (MC100EP210S).

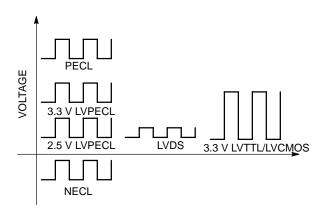


Figure 1. Comparison of Output Voltage Levels Standards (Figure not to Scale)

LVDS require a $100\,\Omega$ load resistor between the differential outputs to generate the Differential Output Voltage (V_{OD}) with a maximum current of 2.5 mA flowing through the load resistor. This load resistor will terminate the $50\,\Omega$ controlled characteristic impedance line, which prevent reflections and reduces unwanted electromagnetic emission (Figure 2).

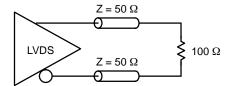


Figure 2. LVDS Output Definition

LVDS receivers require 200 mV minimum input swing within the input voltage range of 0 V to 2.4 V and can tolerate a minimum of ± 1.0 V ground shift between the driver's ground and the receiver's ground, since LVDS receivers have a typical driver offset voltage of 1.2 V. The common mode range of the LVDS receiver is 0.2 V to 2.2 V, and the recommended LVDS receiver input voltage range is from 0 V to 2.4 V. Common mode range of LVDS is similar to the theory of Voltage Input HIGH Common Mode Range (VIHCMR) of ECL devices.

Currently more LVDS standards are being developed as LVDS technology gains in popularity.

AN1568/D

BLVDS

National Semiconductor developed Bus LVDS (BLVDS) for multipoint applications. This standard is targeted at heavily loaded back planes, which reduces the impedance of the transmission line by 50% or more. By providing increased drive current, the double termination seen by the driver will be compensated.

M-L VDS

TIA TR30.2 standards group is developing another multipoint LVDS application called Multipoint LVDS (M-LVDS). The maximum data rate is 500 Mbps.

GLVDS and SLVS

Ground referenced LVDS (GLVDS) is similar to LVDS except the driver output voltage offset is nearer to ground. The advantage of GLVDS is the use of very low power supply voltages (0.5 V).

Similar standard to GLVDS is SLVS (Scalable Low-Voltage Signaling for 400 mV) by JEDEC. The interface is terminated to ground with 400mV swing and a minimum supply voltage of 0.8 V.

LVDM

Texas Instruments developed LVDM, which is designed for double 100 Ω -terminated applications. The driver's output current is two times the standard LVDS, thus producing LVDS characteristic levels.

Table 2. LVDS LEVELS

			DS ication		/DS ication	M-L' Specifi		GL\ Specifi			DM ication		
Symbol	Parameter	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit	Condition
Transmit	ter			•	•	•			•	•	•	•	
V _{PP}	Output Differential Voltage	250	400	240	500	480	650	150	500	247	454	mV	
Vos	Output Offset Voltage	1125	1275	1225	1375	300	2100	75	250	1.125		mV	
R _L	Output Offset Voltage	10	00	27	50	5	0	Interna	l To Rx	5	0	W	
I _{OD}	Output Differen- tial Current	2.5	4.5	9	17	9	13	Adjus	table	6		mA	
Receiver				•	•	•				•	•		
	Input Voltage Range	0	2400	0	2400	-1000	3800	-500	1000	0	2400	MV	V _{gpd} < 950 mV (Note 2)
	Differential HIGH Input Threshold		+100		+100		+50		+100		+100	mV	V _{gpd} < 950 mV (Note 2)
	Differential LOW Input Threshold	-100		-100		-50		-100		-100		mV	V _{gpd} < 950 mV (Note 2)

^{2.} Vgpd is the voltage of Ground Potential Delta across or between boards.

Interfacing

Common mode range inputs are capable of processing differential signals with 150 mV to 400 mV amplitude. The ECL input processes signals up to 1.0 V amplitude. The DC voltage levels should be within the voltage input HIGH common mode range (V_{IHCMR}).

To interface between these 2 voltage levels, capacitive coupling can be used. Only clock or coded signals should be capacitively coupled. A capacitive coupling of NRZ signals will cause problems, which can require a passive or active interfacing.

Capacitive Coupling LVDS to ECL

Capacitive Coupling LVDS to ECL Using V BB

Several ECL devices provide a V_{BB} ($V_{BB} \approx V_{CC} - 1.3V$) reference voltage. It can be used for differential capacitive coupling. The V_{BB} needs to be decoupled to GND via a 10 nF capacitor. (Figure 3)

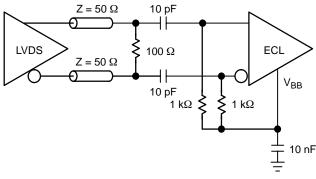


Figure 3. Capacitive Coupling LVDS to ECL Using V_{BB}

Capacitive Coupling LVDS to ECL with External Biasing

If V_{BB} reference voltage is not available, equivalent DC voltage can be generated using a resistor divider network. The resistor values depend on V_{CC} and V_{EE} voltages (Table 3). Stability is enhanced during null signal conditions if a 50 mV differential voltage is maintained between the divider networks. (Figure 4)

Table 3. Examples:

V _{CC} = GND	$V_{EE} = -5.0 \text{ V}$	$R1 = 1.2 \text{ k}\Omega$	$R2 = 3.4 \text{ k}\Omega$
V _{CC} = GND	$V_{EE} = -3.3 \text{ V}$	R1 = 680 Ω	$R2 = 1.0 \text{ k}\Omega$
V _{CC} = GND	V _{EE} = -2.5 V	R1 = 100 Ω	R2 = 90 Ω

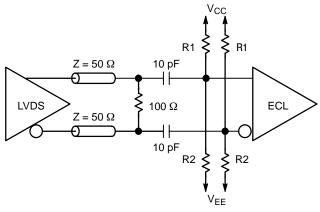


Figure 4. Capacitive Coupling LVDS to ECL with External Biasing

In the layout for both interfaces, the resistors and the capacitors should be located as close as possible to the ECL input to insure reduced reflection and increased signal integrity.

Capacitive Coupling ECL to LVDS

The ECL output requires a DC current path to V_{EE} ; therefore, the pulldown termination resistors, R_T , are connected to V_{EE} . The Thevenin resistor pair represent the termination of the transmission line $Z=R1 \parallel R2$ and generates an appropriate DC offset level of 1.2 V. (Figure 5)

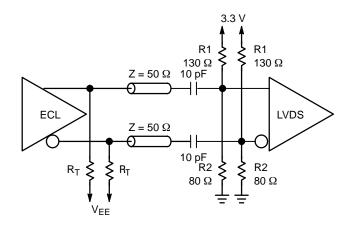


Figure 5. Capacitive Coupling ECL to LVDS

An example of capacitive coupled LVPECL (ECLinPS $Plus^{TM}$ Device) to LVDS is shown below. (Figure 6)

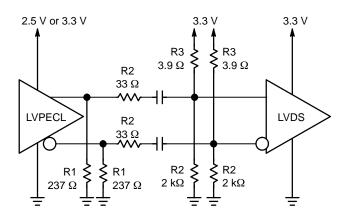


Figure 6. Capacitive Coupling LVPECL to LVDS

Capacitive Coupling ECL to LVDS Using V_{OS} Reference Voltage

Some LVDS devices supply offset reference voltage (V_{OS}) , which can be used for capacitive coupling. When the transmission line is very short, a parallel termination should be used and placed as close as possible to the coupling capacitors. (Figure 7)

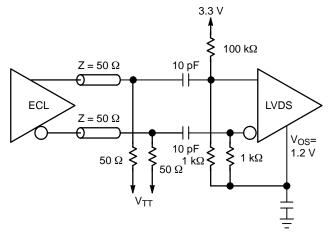


Figure 7. Capacitive Coupling ECL to LVDS Using V_{OS} Reference Voltage

Interfacing from LVPECL to LVDS

The DC output level of LVPECL is more positive than the input range of LVDS, so all ECL devices will require a pulldown resistor. The pulldown resistors in a Thevenin parallel termination or pulldown resistors to GND can be split up into a resistor divider network to generate LVDS DC levels.

Dependent on the application, one of the following interfaces should be used:

Interfacing from LVPECL to LVDS using Thevenin Parallel Termination

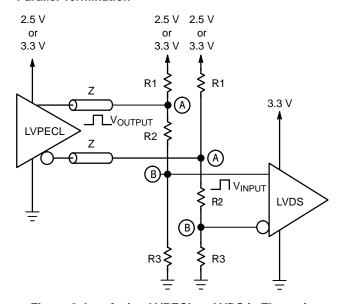


Figure 8. Interfacing LVPECL to LVDS in Thevenin Parallel Termination

The Thevenin parallel resistors terminate the transmission line Z near the receiver (Figure 8). Instead of a resistor to V_{EE} , a resistive path to V_{CC} and to V_{EE} (GND) generates the termination of the transmission line. In transmission line theory these resistors are in parallel for high frequency signals. They match the line characteristic impedance (eq. 1).

$$\frac{1}{\frac{1}{R1} + \frac{1}{R2 + R3}} = Z$$
 (eq. 1)

The DC condition for point A is $V_{CC} - 2.0 \text{ V}$ (eq. 2). The DC levels at the LVDS input (B) are located within the LVDS input common mode range (eq. 3).

A:
$$\frac{R2 + R3}{R1 + R2 + R3} = \frac{VCC - 2V}{VCC}$$
 (eq. 2)

B:
$$\frac{R3}{R1 + R2 + R3} = \frac{V_{IL}}{V_{CC}}$$
 (eq. 3)

The amplitude at the LVDS input is decreased dependent on R2 and R3 resister values.

$$V_{Input} = \frac{R3}{R2 + R3} \cdot V_{Output}$$
 (eq. 4)

$$V_{IH} < 2.0 V (2.4 V)$$
 (eq. 5)

$$VIL > 0 V$$
 (eq. 6)

Examples:

For 50 Ω controlled impedance, the resistor values for 2.5 V LVPECL and 3.3 V LVPECL converted to LVDS voltage levels are as follows:

2.5 V LVPECL System

Resistor Values are:

$$R1 = 63~\Omega$$

$$R2 = 125 \Omega$$

$$R3 = 125 \Omega$$

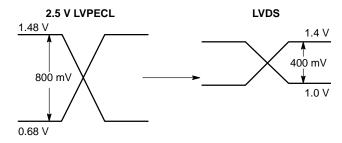


Figure 9. +2.5 V LVPECL to LVDS Voltages Levels

3.3 V LVPECL System

Resistor Values are:

 $R1 = 83 \Omega$

 $R2 = 63 \Omega$ $R3 = 63 \Omega$

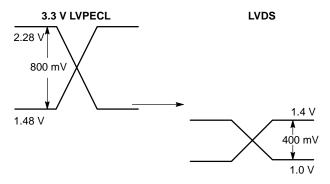


Figure 10. +3.3 V LVPECL to LVDS Voltages Levels

For any other controlled impedance line, the calculation of the resistive divider network is done according to Eq. 1, Eq. 2, and Eq. 3.

Interfacing from LVDS to LVPECL

The input common mode range of the low voltage ECL line receivers are wide enough to process LVDS signals. (Figure 11)

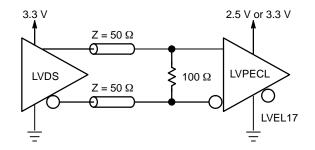


Figure 11. Interfacing LVDS to LVPECL

This direct interface is possible for all ECL devices with sufficiently low minimum differential common mode range inputs. A differentially operated receiver's V_{IHCMR} minimum must be 1.2 V or less (see device data sheet).

The following devices are LVDS input compatible:

	_		
EP14	LVEL11	LVEL51	SG11
EP809	LVEL13	LVEL56	SG14
LVEP11	LVEL14	LVEL92	SG16
LVEP14	LVEL16	EL13	SG16VS
LVEP16	LVEL17	EL14	SG53A
LVEP34	LVEL29	EL17	SG86A
LVEP111	LVEL32	EL29	
LVEP210	LVEL33	EL39	
LVEP210S	LVEL37	EL56	
LVE222	LVEL39	EL91	
LVEL05	LVEL40		

Interfacing from PECL to LVDS

Interfacing from PECL to LVDS Using Thevenin Parallel Termination

As described for LVPECL, to interface from PECL to LVDS a Thevenin Parallel Termination is used.

Near the receiver, a +5.0 V power supply connection is required. (Figure 12)

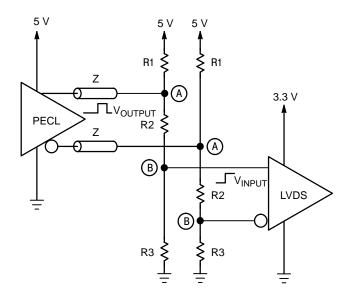


Figure 12. Interfacing from PECL to LVDS Using Thevenin Equivalent Parallel Termination

$$\frac{1}{\frac{1}{R1} + \frac{1}{R2 + R3}} = Z$$
 (eq. 7)

The DC termination level at point A is V_{CC} – 2.0 V. The DC level of input B should be within the LVDS input common mode range.

A:
$$\frac{R2 + R3}{R1 + R2 + R3} = \frac{VCC - 2V}{VCC}$$
 (eq. 8)

B:
$$\frac{R3}{R1 + R2 + R3} = \frac{V_{IL}}{V_{CC}}$$
 (eq. 9)

The LVDS input voltage amplitude (V_{Input}) is decreased from V_{Output} , which is dependent on R2 and R3 resister values:

$$V_{Input} = \frac{R3}{R2 + R3} \cdot V_{Output}$$
 (eq. 10)

$$V_{IH} < 2.0 V (2.4 V)$$
 (eq. 11)

$$V_{IL} > 0 V$$
 (eq. 12)

Calculations may give non-standard resistor values. When choosing resistors off the shelf, keep in mind to avoid a cutoff condition under worst-case condition.

If a 50 Ω controlled impedance line is used the following resistor values are useful:

Examples:

For 50 Ω controlled impedance, the resistor values for 5.0 V PECL converted to LVDS voltage levels are as follows:

5.0 V PECL System

Resistor Values are:

 $R1 = 125 \Omega$

 $R2 = 42 \Omega$

 $R3 = 42 \Omega$

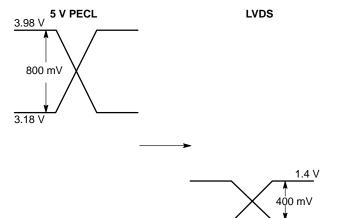


Figure 13. Interfacing from PECL to LVDS Using Thevenin Equivalent Parallel Termination

For any other controlled impedance line the calculation of the resistive divider network is done according to Eq. 7, Eq. 8 and Eq. 9.

Interfacing from +3.3 V LVDS to +5.0 V PECL

To translate LVDS signals to PECL a differential ECL device with extended common mode range inputs (e.g., MC100EL17) can be used to process and translate LVDS signals when supplied with 5.0 V \pm 5% supply voltage. (See Figure 14)

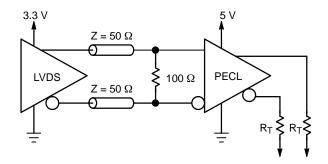


Figure 14. Interfacing LVDS to PECL

AN1568/D

Interfacing Between NECL to LVDS

ON Semiconductor has developed level translators to interface between the different voltage levels. The

MC100EP90 translates from negative supplied ECL to LVPECL. The interface from LVPECL to LVDS inputs is described above. (Figure 15)

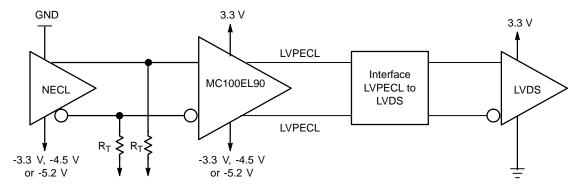


Figure 15. Interfacing from NECL to LVDS

To interface from LVDS to negative supplied ECL the common mode range (V_{IHCMR}) of the MC100LVEL91 for -3.3~V supply and the MC100EL91 for -4.5~V/-5.2~V supply is wide enough to process LVDS signals. (See Figure 16)

If V_{CC} = +5 V \pm 5% supply and a V_{EE} = -5.2 V \pm 5% supply is available the MC10E1651 can be used.

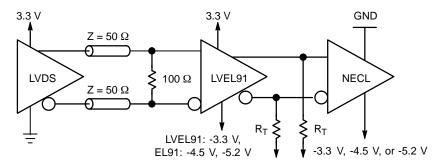


Figure 16. Interfacing from LVDS to NECL

ECLinPS Plus is a trademark of Semiconductor Components Industries, LLC (SCILLC).

ON Semiconductor and are registered trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. "Typical" parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the SCILLC product could create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that SCILLC was negligent regarding the design or manufacture of the part. SCILLC is an Equal Opportunity/Affirmative Action Employer.

PUBLICATION ORDERING INFORMATION

Literature Fulfillment:

Literature Distribution Center for ON Semiconductor P.O. Box 5163, Denver, Colorado 80217 USA

Phone: 303-675-2175 or 800-344-3860 Toll Free USA/Canada **Fax**: 303-675-2176 or 800-344-3867 Toll Free USA/Canada

Email: ONlit@hibbertco.com

N. American Technical Support: 800-282-9855 Toll Free USA/Canada

JAPAN: ON Semiconductor, Japan Customer Focus Center 2-9-1 Kamimeguro, Meguro-ku, Tokyo, Japan 153-0051

Phone: 81-3-5773-3850 Email: r14525@onsemi.com

ON Semiconductor Website: http://onsemi.com

For additional information, please contact your local Sales Representative.